ECE-205 : Dynamical Systems

Homework #6

Due: Friday January 28 at noon

1) In this problem we will derive some useful properties of Laplace transforms starting from the basic relationship

$$X(s) = \int_{0^{-}}^{\infty} x(t)e^{-st}dt$$

a) Let's assume x(t) is a causal signal (it is zero for t < 0). We can then write x(t) = x(t)u(t) to emphasize the fact that x(t) is zero before time zero. If there is a delay in the signal and it starts at time t_0 , then we can write the signal as $x(t-t_0) = x(t-t_0)u(t-t_0)$ to emphasize the fact that the signal is zero before time t_0 .

Using the definition of the Laplace transform and a simple change of variable in the integral, show that $\mathcal{L}\{x(t-t_0)u(t-t_0)\}=X(s)e^{-st_0}$

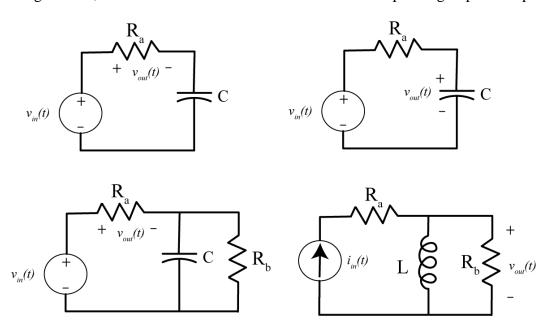
b) Using the results from part **a**, determine the inverse Laplace transform of $X(s) = \frac{e^{-3s}}{(s+2)(s+4)}$

Answer:
$$x(t) = \frac{1}{2} \left[e^{-2(t-3)} - e^{-4(t-3)} \right] u(t-3)$$

- c) Starting from the definition of the Laplace transform, show that $\mathcal{L}\{tx(t)\} = -\frac{dX(s)}{ds}$.
- **d)** Using the result from part c, and the transform pair $x(t) = e^{-at}u(t) \leftrightarrow X(s) = \frac{1}{s+a}$, and some simple calculus, show that

$$\mathcal{L}\left\{te^{-at}u(t)\right\} = \frac{1}{(s+a)^2}, \,\mathcal{L}\left\{t^2e^{-at}u(t)\right\} = \frac{2}{(s+a)^3}, \,\mathcal{L}\left\{t^3e^{-at}u(t)\right\} = \frac{6}{(s+a)^4}$$

2) For the following circuits, determine the transfer function and the corresponding impulse responses.



Scrambled Answers:

$$h(t) = R_b \delta(t) - \frac{R_b^2}{L} e^{-tR_b/L} u(t), h(t) = \delta(t) - \frac{1}{R_a C} e^{-t/R_a C} u(t), h(t) = \frac{1}{R_a C} e^{-t/R_a C} u(t), h(t) = [\delta(t) - \frac{1}{R_a C} e^{-t\frac{R_a + R_b}{CR_a R_b}}] u(t)$$

3) For the following impulse responses and inputs, compute the system output using transfer functions.

a)
$$h(t) = e^{-t}u(t)$$
, $x(t) = u(t)$ **b**) $h(t) = e^{-2t}u(t)$, $x(t) = \delta(t)$ **c**) $h(t) = e^{-2(t-1)}u(t-1)$, $x(t) = e^{-2t}u(t)$

d)
$$h(t) = e^{-t}u(t), x(t) = (t-1)u(t-1)$$
 e) $h(t) = e^{-2t}u(t), x(t) = u(t) - u(t-1)$

f)
$$h(t) = e^{-2(t-1)}u(t-1), x(t) = te^{-3t}u(t)$$

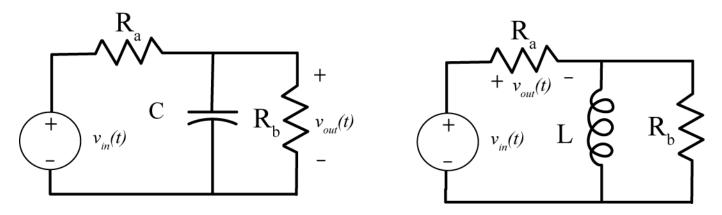
Scrambled Answers:

$$y(t) = \frac{1}{2} \left[1 - e^{-2t} \right] u(t) - \frac{1}{2} \left[1 - e^{-2(t-1)} \right] u(t-1), \ y(t) = \left(1 - e^{-t} \right) u(t), \ y(t) = (t-1)e^{-2(t-1)} u(t-1)$$

$$y(t) = e^{-2t} u(t), \ y(t) = \left[-1 + (t-1) + e^{-(t-1)} \right] u(t-1),$$

$$y(t) = \left[e^{-2(t-1)} - e^{-3(t-1)} - (t-1)e^{-3(t-1)} \right] u(t-1)$$

4) For the following circuits, determine and expression for the output $V_{out}(s)$ in terms of the ZSR and ZIR. Do not assume the initial conditions are zero. Also determine the system transfer function.



Answers:

$$V_{out}(s) = \left[\frac{R_b}{R_a R_b C s + R_a + R_b}\right] V_{in}(s) + \left[\frac{R_a R_b C}{R_a R_b C s + R_a + R_b}\right] v(0^{-})$$

$$V_{out}(s) = \left[\frac{R_a (R_b + L s)}{(R_a + R_b) L s + R_a R_b}\right] V_{in}(s) + \left[\frac{R_a R_b L}{(R_a + R_b) L s + R_a R_b}\right] i(0^{-})$$

5) For the following transfer functions

$$H(s) = \frac{2}{s^2 + 2s + 2} \quad H(s) = \frac{3}{s^2 + 4s + 6} H(s) = \frac{5}{s^2 + 6s + 10}$$

$$H(s) = \frac{4}{s^2 - 4s + 7} \quad H(s) = \frac{1}{s^2 + 4}$$

By computing the inverse Laplace transform show that the step responses are given by

$$y(t) = \left[1 - e^{-t}\cos(t) - e^{-t}\sin(t)\right]u(t) \quad y(t) = \left[\frac{1}{2} - \frac{1}{\sqrt{2}}e^{-2t}\sin(\sqrt{2}t) - \frac{1}{2}e^{-2t}\cos(\sqrt{2}t)\right]u(t)$$

$$y(t) = \left[\frac{1}{2} - \frac{3}{2}e^{-3t}\sin(t) - \frac{1}{2}e^{-3t}\cos(t)\right]u(t) \quad y(t) = \left[\frac{4}{7} + \frac{8\sqrt{3}}{21}e^{2t}\sin(\sqrt{3}t) - \frac{4}{7}e^{2t}\cos(\sqrt{3}t)\right]u(t)$$

$$y(t) = \left[\frac{1}{4} - \frac{1}{4}\cos(2t)\right]u(t)$$