## ECE-320: Linear Control Systems Homework 9

Due: Thursday February 13 at the beginning of class

1) Consider the continuous-time plant with transfer function

$$G_p(s) = \frac{1}{(s+1)(s+2)}$$

We want to determine the discrete-time equivalent to this plant,  $G_p(z)$ , by assuming a zero order hold is placed before the continuous-time plant to convert the discrete-time control signal to a continuous time control signal.

Show that if we assume a sampling interval of T, the equivalent discrete-time plant is

$$G_p(z) = \frac{z(0.5 - e^{-T} + 0.5e^{-2T}) + (0.5e^{-T} - e^{-2T} + 0.5e^{-3T})}{(z - e^{-T})(z - e^{-2T})}$$

Note that we have poles were we expect them to be, but we have introduced a zero in going from the continuous time system to the discrete-time system.

2) Consider the discrete-time state variable model  $\underline{x}(k+1) = G(T)\underline{x}(k) + H(T)u(k)$ 

where the explicit dependence of *G* and *H* on the sampling time *T* has been emphasized. Here  $G(T) = e^{AT}$ 

$$H(T) = \int_0^T e^{A\lambda} d\lambda B$$

**a**) Show that if A is invertible, we can write  $H(T) = [e^{AT} - I]A^{-1}B$ 

**b**) Show that if A is invertible and T is small we can write the state model as

$$\underline{x}(k+1) = [I + AT]\underline{x}(k) + BTu(k)$$

c) Show that if we use the approximation

$$\underline{\dot{x}}(t) \approx \frac{\underline{x}((k+1)T) - \underline{x}(kT)}{T} = Ax(kT) + Bu(kT)$$

we get the same answer as in part **b**, but using this approximation we do not need to assume *A* is invertible.

**d**) Show that if we use two terms in the approximation for  $e^{AT}$  (and no assumptions about A being invertible), we can write the state equations as

$$\underline{x}(k+1) = \left[I + AT\right]\underline{x}(k) + \left[IT + \frac{1}{2}AT^2\right]Bu(k)$$

3) For the state variable system

$$\underline{\dot{x}}(t) = \begin{bmatrix} 1 & -1 \\ 2 & 4 \end{bmatrix} \underline{x}(t) + \begin{bmatrix} 1 \\ 1 \end{bmatrix} u(t)$$

a) Show that

$$e^{AT} = \begin{bmatrix} 2e^{2T} - e^{3T} & e^{2T} - e^{3T} \\ 2e^{3T} - 2e^{2T} & 2e^{3T} - e^{2T} \end{bmatrix}$$

b) Derive the equivalent ZOH discrete-time system

$$\underline{x}(k+1) = G\underline{x}(k) + Hu(k)$$

for T = 0.1 (integrate each entry in the matrix  $e^{A\lambda}$  separately.) Compare your answer with that given by Matlab's **c2d** command, [G,H] = c2d(A,B,T).

**4**) For the matrix 
$$A = \begin{bmatrix} 1 & 0 \\ 1 & 1 \end{bmatrix}$$
 show that  $e^{At} = \begin{bmatrix} e^t & 0 \\ te^t & e^t \end{bmatrix}$ 

**5**) Consider the discrete-time state variable model  $\underline{x}(k+1) = G\underline{x}(k) + Hu(k)$  with the initial state x(0) = 0. Let

$$G = \begin{bmatrix} 0 & 1 \\ 1 & 0 \end{bmatrix}, H = \begin{bmatrix} 1 \\ 1 \end{bmatrix}, C = \begin{bmatrix} 1 & 0 \end{bmatrix}, D = 0$$

a) Determine the corresponding transfer function for the system.

b) Using state variable feedback with  $u(k) = G_{pf}r(k) - Kx(k)$  show that the transfer function is given by

$$F(z) = \frac{Y(z)}{R(z)} = C(zI - \tilde{G})^{-1}\tilde{H} = \frac{G_{pf}(z+1)}{(z+k_1)(z+k_2) - (k_1-1)(k_2-1)}$$

c) Show that if  $G_{pf} = 1$  and  $k_1 = k_2 = 0$ , the transfer function reduces to that found in part **a**.

d) Is the system controllable? That is, is it possible to find  $k_1$  and  $k_2$  to place the poles of the closed loop system where ever we want? For example, can we make both poles be zero?

6) Consider the discrete-time state variable model

$$\underline{x}(k+1) = G\underline{x}(k) + Hu(k)$$

with the initial state x(0) = 0. Let

$$G = \begin{bmatrix} 1 & 0 \\ 1 & 1 \end{bmatrix}, H = \begin{bmatrix} 0 \\ 1 \end{bmatrix}, C = \begin{bmatrix} 0 & 1 \end{bmatrix}, D = 0$$

a) Determine the corresponding transfer function for the system.

b) Using state variable feedback with  $u(k) = G_{vf}r(k) - Kx(k)$  show that the transfer function is given by

$$F(z) = \frac{Y(z)}{R(z)} = \frac{G_{pf}(z-1)}{(z-1)(z+k_2-1)}$$

c) Show that if  $G_{pf} = 1$  and  $k_1 = k_2 = 0$ , the transfer function reduces to that found in part **a**.

d) Is the system controllable?

6) Consider the discrete-time state variable model

$$\underline{x}(k+1) = G\underline{x}(k) + Hu(k)$$

with the initial state x(0) = 0. Let

$$G = \begin{bmatrix} 0 & 1 \\ 1 & 1 \end{bmatrix}, H = \begin{bmatrix} 0 \\ 1 \end{bmatrix}, C = \begin{bmatrix} 1 & 0 \end{bmatrix}, D = 0$$

a) Determine the corresponding transfer function for the system.

b) Using state variable feedback with  $u(k) = G_{vf}r(k) - Kx(k)$  show that transfer function is given by

$$F(z) = \frac{Y(z)}{R(z)} = \frac{G_{pf}}{z^2 + (k_2 - 1)z + (k_1 - 1)}$$

c) Show that if  $G_{pf} = 1$  and  $k_1 = k_2 = 0$ , the transfer function reduces to that found in part **a**.

d) Is it possible to find  $k_1$  and  $k_2$  to place the poles of the closed loop system where ever we want? For example, can we make both poles be zero? If we want the poles to be at  $p_1$  and  $p_2$  show that  $k_2 = 1 - (p_1 + p_2)$  and  $k_1 = 1 + p_1 p_2$ .