

Name Solution 5 Mailbox _____

ECE-320 Linear Control Systems

Winter 2012, Exam 2

No calculators or computers allowed.

Problem 1	_____	/20
Problem 2	_____	/20
Problem 3	_____	/12
Problem 4	_____	/12
Problems 5-16	_____	/36
Total	_____	/100

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1) (20 points) For impulse response $h(n) = \left(\frac{1}{2}\right)^{n+1} u(n+2)$ and input $x(n) = \left(\frac{1}{4}\right)^{n-1} u(n-2)$,

determine the system output by evaluating the convolution sum $y(n) = \sum_{k=-\infty}^{\infty} h(n-k)x(k)$

Note: you do not have to simplify your answer, but you must remove all sums and include a unit step function of some sort.

$$y(n) = \sum_{k=-\infty}^{\infty} \left(\frac{1}{2}\right)^{n-k+1} u(n-k+2) \left(\frac{1}{4}\right)^{k-1} u(k-2)$$

$$= \sum_{k=2}^{k=n+2} \left(\frac{1}{2}\right)^{n+1} \left(\frac{1}{2}\right)^{-k} \left(\frac{1}{4}\right)^k \left(\frac{1}{4}\right)^{-1}$$

$$= \left(\frac{1}{2}\right)^{n+1} 4 \sum_{k=2}^{k=n+2} \left(\frac{1}{2}\right)^k \quad \text{let } l = k-2$$

$$l+2 = k$$

$$= \left(\frac{1}{2}\right)^{n+1} 4 \sum_{l=0}^{l=n} \left(\frac{1}{2}\right)^{l+2} = \left(\frac{1}{2}\right)^{n+1} 4 \left(\frac{1}{2}\right)^2 \sum_{l=0}^{l=n} \left(\frac{1}{2}\right)^l$$

$$= \left(\frac{1}{2}\right)^{n+1} \left[\frac{1 - \left(\frac{1}{2}\right)^{n+1}}{1 - \frac{1}{2}} \right] = \left(\frac{1}{2}\right)^n \left[1 - \left(\frac{1}{2}\right)^{n+1} \right]$$

$$y(n) = \left(\frac{1}{2}\right)^n \left[1 - \left(\frac{1}{2}\right)^{n+1} \right] u(n)$$

$$n-k+2 \geq 0$$

$$n+2 \geq k$$

$$k-2 \geq 0$$

$$k \geq 2$$

$$n+2 \geq k \geq 2$$

$$n+2 \geq 2$$

$$n \geq 0$$

2) (20 points) For impulse response $h(n) = \left(\frac{1}{2}\right)^{n+1} u(n-1)$ and input $x(n) = \left(\frac{1}{3}\right)^{n+1} u(n-2)$,

a) determine the z-transform of $h(n)$, $H(z)$

b) determine the z-transform of $x(n)$, $X(z)$

c) determine $y(n)$

Hint: Assume $Y(z) = z^{-2}G(z)$, determine $g(n)$ and then $y(n)$

$$a) h(n) = \left(\frac{1}{2}\right)^{n+1} u(n-1) = \left(\frac{1}{2}\right)^{n-1} \left(\frac{1}{2}\right)^2 u(n-1) \Rightarrow H(z) = \frac{1}{4} z^{-1} \frac{z}{z-\frac{1}{2}}$$

$$H(z) = \frac{\frac{1}{4}}{z-\frac{1}{2}}$$

$$b) x(n) = \left(\frac{1}{3}\right)^{n+1} u(n-2) = \left(\frac{1}{3}\right)^{n-2} \left(\frac{1}{3}\right)^3 u(n-2) \Rightarrow X(z) = \frac{1}{27} z^{-2} \frac{z}{z-\frac{1}{3}}$$

$$X(z) = \frac{\frac{1}{27} z^{-1}}{z-\frac{1}{3}}$$

$$c) Y(z) = H(z) X(z) = \frac{\frac{1}{4} \frac{1}{27} z^{-1}}{(z-\frac{1}{2})(z-\frac{1}{3})} \quad G(z) = z^2 Y(z) = \frac{\frac{1}{4} \frac{1}{27} z}{(z-\frac{1}{2})(z-\frac{1}{3})}$$

$$\frac{G(z)}{z} = \frac{\frac{1}{4} \frac{1}{27}}{(z-\frac{1}{2})(z-\frac{1}{3})} = \frac{A}{z-\frac{1}{2}} + \frac{B}{z-\frac{1}{3}}$$

$$A = \frac{\frac{1}{4} \cdot \frac{1}{27}}{\frac{1}{6}} = \frac{6}{4 \cdot 27} = \frac{1}{18}$$

$$B = \frac{\frac{1}{4} \cdot \frac{1}{27}}{-\frac{1}{6}} = -\frac{6}{4 \cdot 27} = -\frac{1}{18}$$

$$g(n) = \frac{1}{18} \left(\frac{1}{2}\right)^n u(n) - \frac{1}{18} \left(\frac{1}{3}\right)^n u(n)$$

$$y(n) = g(n-2) = \frac{1}{18} \left[\left(\frac{1}{2}\right)^{n-2} - \left(\frac{1}{3}\right)^{n-2} \right] u(n-2) = y(n)$$

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3) (12 points) Consider the continuous-time plant with transfer function

$$G_p(s) = \frac{1}{s}$$

We want to determine the discrete-time equivalent to this plant, $G_p(z)$, by assuming a zero order hold is placed before the continuous-time plant to convert the discrete-time control signal to a continuous time control signal.

$$G_p(z) = (1 - z^{-1}) Z \left\{ \frac{1}{s^2} \right\} \quad Z \left\{ \frac{1}{s^2} \right\} = \frac{Tz}{(z-1)^2}$$

$$G_p(z) = (1 - z^{-1}) \frac{Tz}{(z-1)^2} = \left(\frac{z-1}{z} \right) \frac{Tz}{(z-1)^2} = \frac{T}{z-1} = G_p(z)$$

4) (12 points) In this problem we derive discrete-time equivalents to continuous-time controllers. You must show work or you will receive no points (you cannot just write down the answers!)

a) Consider the continuous-time integral controller, $G_c(s) = k_i \left(\frac{1}{s} \right)$. By determining the discrete-time equivalent to the function $\frac{1}{s}$, find the equivalent discrete-time integral controller.

$$z \left\{ \frac{1}{s} \right\} = \frac{z}{z-1} = \frac{1}{1-z^{-1}} \quad G_c(z) = \frac{k_i}{1-z^{-1}}$$

b) Consider the continuous-time derivative controller, $G_c(s) = k_d s = \frac{U(s)}{E(s)}$. Using the continuous-time approximation $u(t) = k_d \left[\frac{e(t) - e(t-T)}{T} \right]$, take z-transforms to find the equivalent discrete-time derivative controller.

$$u(nT) = k_d \left[\frac{e(nT) - e((n-1)T)}{T} \right] \quad U(z) = \frac{k_d}{T} [E(z) - z^{-1}E(z)]$$

$$\frac{U(z)}{E(z)} = G_c(z) = \frac{k_d}{T} (1 - z^{-1})$$

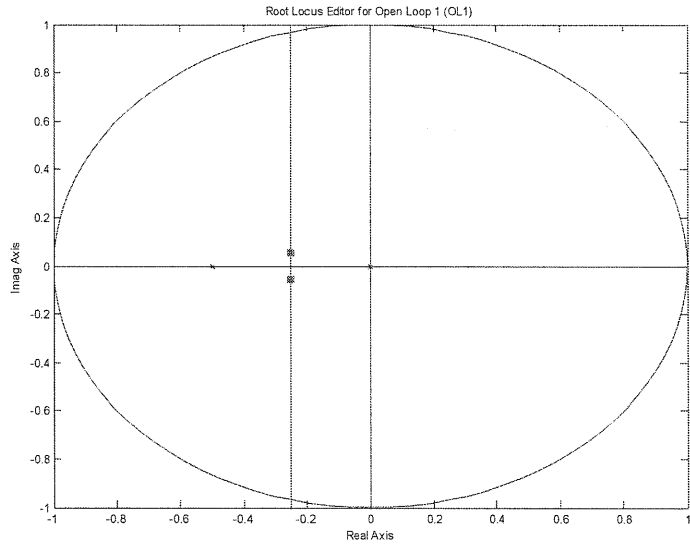
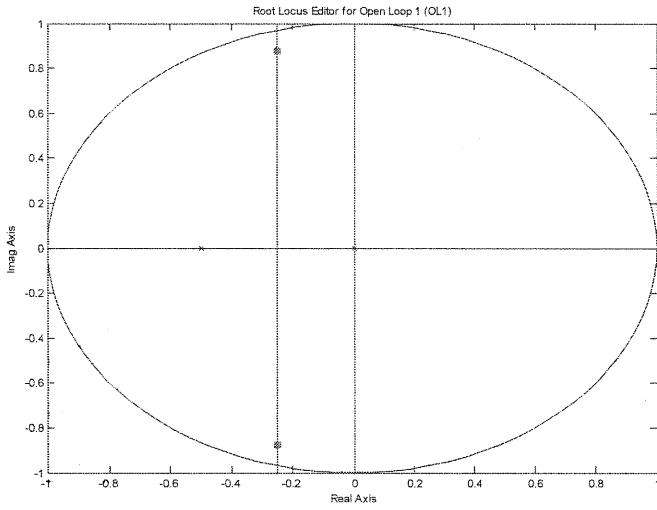
c) Assuming we use *sisotool* to determine the PI controller $G_c(z) = \frac{7(z-5)}{z-1} = k_p + \frac{k_i}{1-z^{-1}}$. Determine k_p and k_i .

$$\frac{7(z-5)}{z-1} = \frac{7z-5}{z-1} = k_p + \frac{k_i}{1-z^{-1}} = k_p + \frac{k_i z}{z-1}$$

$$= \frac{k_p z - k_p + k_i z}{z-1}$$

$$7z-5 = (k_p + k_i)z - k_p \quad \boxed{k_p = 5 \quad k_i = 2}$$

Problems 5 and 6 refer to the following two root locus plot for a discrete-time system



5) For which system is the settling time likely to be smallest?

- a) The system on the left **b) the system on the right** c) the settling time will be the same

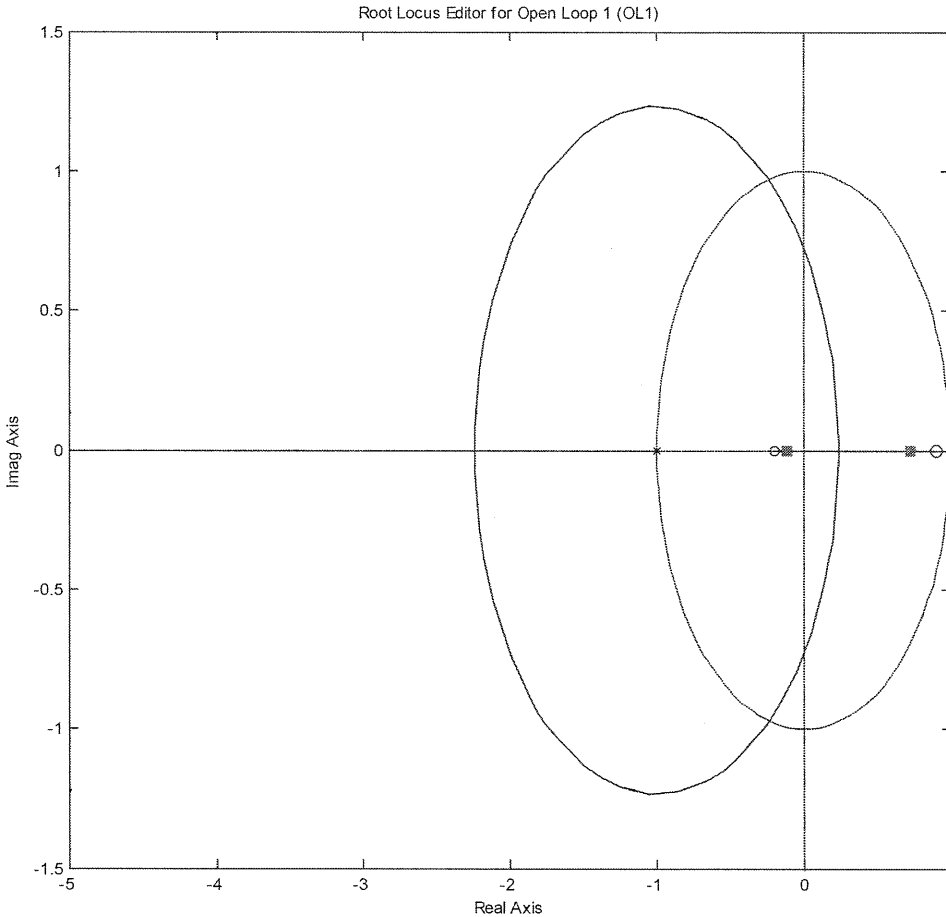
poles close to origin

6) Is this a type 1 system?

- a) yes **b) no** c) not enough information

no poles at z = 1

Problems 7-9 refer to the following root locus plot for a discrete-time system



7) With the closed loop pole locations shown in the figure, is the closed loop system stable?

- a) yes b) no c) not enough information *all poles inside unit circle*

8) Is there any value of k for which the closed loop system is stable?

- a) yes b) no c) not enough information

9) Is this a type one system?

- a) yes b) no c) not enough information
no poles at $z=1$

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10) Is the following system *controllable*? $G(s) = \frac{G_{pf}}{(s - k_1 k_2)^2}$

a) Yes b) No c) impossible to determine

11) Is the following system *controllable*? $G(s) = \frac{8G_{pf}}{s^2 + 12s + (k_1 + k_2 + 20)}$

a) Yes b) No c) impossible to determine

12) Is the following system *controllable*? $G(s) = \frac{G_{pf}}{s^2 + (k_2 + k_1 - 1)s + (k_2 + 2)}$

a) Yes b) No c) impossible to determine

13) Consider a plant that is unstable but is a controllable system. Is it possible to use state variable feedback to make this system stable?

a) Yes b) No

14) Is it possible for a system with state variable feedback to change the zeros of the plant (other than by pole-zero cancellation)?

a) Yes b) No

15) Is it possible for a system with state variable feedback to introduce zeros into the closed loop system?

a) Yes b) No

16) If a plant has n poles, then a system with state variable feedback with no pole-zero cancellations will have

a) more than n poles b) less than n poles c) n poles d) it is not possible to tell